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Design and Simulation of 32-Point FFT Using Mixed Radix Algorithm for FPGA Implementation

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ABSTRACT

This paper focus on the development of the fast Fourier transform (FFT), based on Decimation-In-Time (DIT) domain by using Mixed-Radix algorithm (Radix-4 and Radix-8).Fast Fourier transforms, popularly known as FFTs, have become an integral part of any type of digital communication system and a wide variety of approaches have been tried in order to optimize the algorithm for a variety of parameters such as area, time delay, power and speed. In this paper a comparison of area and minimum time delay are drawn between the proposed design of 32 point FFT by using Mixed-Radix algorithmwithRadix-2 algorithm. Here our goal is to implement Mixed Radix 32-point FFT by using hardware language (VHDL). Simulation and synthesis of design is done using Xilinx ISE 14.7. The result shows significant reduction in terms of area complexity and minimum time delay.

Keywords:- Fast Fourier transform (FFT), Discrete Fourier transform (DFT), DIT, Radix-4, 8, VHDL, FPGA.

I. INTRODUCTION

Currently in the field of signal processing for communications, there is a rapid development in FFT algorithms which act as a key in designing a system. This paper explains the implementation and simulation of 32-point FFT using mixed-radix algorithm. Due to radix-4 and radix-8, FFT can accomplish minimum time delay, reduce the area complexity and also achieve cost effective performance with less development time [1].

The FFT is an efficient class of computational algorithms of the DFT. The Discrete Fourier Transform (DFT) is one of the most popular transformations representing discrete signals in the frequency domain. Signals can be transformed between the time and the frequency domain through various transforms [2]. It is necessary to devise efficient algorithms to compute the DFT as the DFT is important component in many an practical applications. FFT algorithms optimize the DFT by eliminating redundant calculations [3]. FFT algorithms eliminate redundant calculations in the computation of the DFT and are therefore much faster .Cooley and Tukey developed FFT algorithm to

reduce the computations of the Discrete Fourier Transform (DFT) from N^2 to N/2 log2N multiplications and N(N-1) to N log2N additions[4]. Several algorithms are developed to reduce the computational complexity, which includes Radix-2, Radix-4, Mixed-Radix, Split-Radix. All these algorithms are developed on one method, that is, Divide and Conquer method [5]. Fast Fourier Transform (FFT) is based on decomposition and breaking the transform into smaller sequences and at last again combining into one transform.

In this design the coding is done in VHDL& their synthesis and simulation is done using Xilinx ISE Design Suite 14.7 [6].

FFT Algorithm: The N-point discrete Fast Fourier Transform(DFT) is defined as [7]:

$$X(k) = \sum_{n=0}^{N-1} x(n) W_N^{nk}$$

$$0 \le n \le N - 1$$
 , $0 \le k \le N - 1$
(1)

Where X(k) is the frequency domain and x(n) is time domain of sequence and W_N^{nk} is called Twiddle factor.

II. RADIX-8 FFT ALGORITHM

In Fig.1the basic butterfly structure for Radix-8 is shown which is used in first stage of FFT block diagram. The numbers of over flow lines indicates the twiddle factor is to be multiplied with the samples [8][11].In this work the number of stages are reduced to 2 since N=4*8 that is; only 2 stages. The following will explains the functioning of Radix-8 FFT Algorithm.

In this Mixed Radix the length of N=32-points is recursively partitions into four parts DFTs of groups of every eighth sample. Due to the presence of eighth sample in every DFT parts for computational the DIT Radix-8 FFT is used. The outputs of these shorter FFTs are reused to compute many outputs, which are greatly reduced the total computational cost. In this Mixed Radix the DFT equation is rearranges into four parts: sums over all groups consists of every eighth discrete-time index n=[0,4,8,.....N-4], n=[1,5,9,...N-3], n=[2,6,10...N-2], n=[3,7,11...N-1] ,(This works only when the FFT length is multiple of eight),further mathematical manipulation shows that the length-N DFT can be computed as the sum of the even-indexed and odd-indexed discrete-time samples, respectively, where all the time samples are multiplied by twiddle factors W_N^{k} , W_N^{2k} , W_N^{3k} .

The following equations illustrate Radix-8 DIT-FFT, which the N-point input sequence splits into four subsequence's, x(4n), x(4n+1), x(4n+2), x(4n+3), n=0,1,... Each subsequence consists a group of eight samples in order to computations these samples the Radix-8 DIT-FFT is used. Equation (1) can be written as follows:

$$X(k) = \sum_{n=0}^{N/4-1} x(4n) W_N^{4nk} + \sum_{n=0}^{N/4-1} x(4n+1) W_N^{(4n+1)k} + \sum_{n=0}^{N/4-1} x(4n+2) W_N^{(4n+2)k}$$

$$+\sum_{n=0}^{N/4-1} x(4n+3) W_N^{(4n+3)k}$$
(2)

$$X(k) = X_1(k) + X_2(k) W_N^{k} + X_3(k) W_N^{2k} + X_4(k) W_N^{3k}$$
(3)

Where
$$X_1(\mathbf{k}) = \sum_{n=0}^{N-1} x(4n) W_N^{4nk}, X_2(\mathbf{k}) = \sum_{n=0}^{N-1} x(4n+1) W_N^{4nk}$$

 $X_3(\mathbf{k}) = \sum_{n=0}^{N/4-1} x(4n+2) W_N^{4nk}, X_4(\mathbf{k}) \sum_{n=0}^{N/4-1} x(4n+3) W_N^{4nk}$



Fig.1.The basic butterfly for radix-8 FFT algorithm

III. RADIX-4 FFT ALGORITHM

In Fig. 2(a) and (b) shows the basic butterfly operation of radix-4 which have four inputs and four outputs; inputs are as x(n), x(n + n/4), x(n + n/2) and x(n + 3n/4) outputs are in normal order X(k). FFTs in which N = r*k, and where the butterflies used in each stage are the same, so called radix-r algorithms.

Radix-4 algorithms have a computational advantages over radix-2 algorithms because one radix-4 butterfly does the work of four radix-2 butterflies, and the radix-4 butterfly requires only three complex multipliers compared to four complex multipliers of four radix-2 butterflies [10].



Fig. 2(a). The basic butterfly for radix-4 FFT algorithm



Fig. 2(b). The basic structure for radix-4 FFT

In this Mixed Radix algorithm for N=32-point the arithmetic operation of radix-4 DIT FFT is defined as-X(k) = $X_1(k)+X_2(k) W_N^{\ k}+X_3(k) W_N^{\ 2k}+X_4(k) W_N^{\ 3k}$ (3)

As shown in Fig.3 the block diagram of Mixed Radix (Radix-4 & 8) decimation in time (DIT) the method used for N=32-points FFT algorithm. In FFT blocks inputs are in digit-reversed order while the outputs are in normal order. Radix-4 decimation in time (DIT) is used in second stage of FFT of N=32-point length. The outputs of first stage of FFT are used as inputs which are processed with radix-4 FFT.





IV. SIMULATION AND RESULTS

The RTL block thus obtained for the decimation in time domain by using Mixed Radix (radix-4 and radix-8) fast Fourier transform algorithm is shown in Fig.4. The internal architecture of the Butterfly Component is shown in Fig.5.



Fig.4. RTL view of 32 Point FFT



Fig.5. Internal Architecture of Butterfly Component

The proposed FFT block of signal length 32 has been simulated and synthesized using the Xilinx and the waveforms obtained after simulating are shown in Fig.6,7,8.

						2,199,999 ps
Name	Value	2,199,995 ps	2,199,996 ps	2,199,997 ps	2,199,998 ps	2,199,999 ps
🖓 cik	1					
🔓 reset	0					
🕨 📑 x0[7:0]	00000101			00000101		
🕨 📑 x1[7:0]	00010001			00010001		
🕨 📑 x2[7:0]	00001000			00001000		
🕨 📑 x3[7:0]	00010101			00010101		
🕨 📑 x4[7:0]	00010010			00010010		
🕨 📑 x5[7:0]	00001001			00001001		
🕨 📑 x6[7:0]	00010100			00010100		
🕨 📑 x7[7:0]	00100000			00100000		
🕨 📑 x8[7:0]	00000110			00000110		
🕨 📑 x9[7:0]	00100001			00100001		
▶ 📑 x10[7:0]	00000001			00000001		
▶ 📑 x11[7:0]	00010011			00010011		
▶ 📑 x12[7:0]	00010101			00010101		
▶ 📑 x13[7:0]	00010111			00010111		
▶ 📑 x14[7:0]	00011000			00011000		
▶ 📑 x15[7:0]	00000111			00000111		
▶ 📑 x16[7:0]	00000011			00000011		
▶ 📑 x17[7:0]	00000111			00000111		
▶ 📑 x18[7:0]	00010000			00010000		
▶ 📑 x19[7:0]	00010010			00010010		
▶ 📑 x20[7:0]	00000010			00000010		
▶ 📑 x21[7:0]	00100011			00100011		
						·
		X1: 2,199,999 ps				

Fig.6. Simulation results of the 32-point FFT

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										2,199,999 ps
Na	me		Value		2,199,995 ps		2,199,996 ps	2,199,997 ps	2,199,998 ps	2,199,999 ps
		x21[7:0]	00100011					00100011		
	I	x22[7:0]	00010111					00010111		
		x23[7:0]	00010010					00010010		
	I	x24[7:0]	00000001					00000001		
	I	x25[7:0]	00000100					00000100		
	I	x26[7:0]	00010110					00010110		
	I	x27[7:0]	00010001					00010001		
	1	x28[7:0]	00000110					00000110		
	I	x29[7:0]	00001001					00001001		
	I	x30[7:0]	00010101					00010101		
	I	x31[7:0]	00010011					00010011		
	0	ro0[52:0]	00000000		0	000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	0000000111100101	
	0	ro8[52:0]	11111111		1	111	1111111111111111111	11111111111111111111111	1111111110110111	
	0	ro16[52:0]	11111111		1	111	1111111111111111111	111111111111111111111111111111111111111	1111111110100101	
	0	ro24[52:0]	11111111		1	111	1111111111111111111	111111111111111111111111111111111111111	1111111110110111	
	0	ro1[51:0]	11111111		1	11	111111111111111111111111111111111111111	1111111111111111111111	1111111111110100	
	0	ro2[51:0]	11111111		1	11	111111111111111111111111111111111111111	11111111111111111111111	111111111001111	
	0	ro3[51:0]	00000000		0	000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	0000000000011110	
		ro4[51:0]	11111111		1	11	1111111111111111111111	1111111111111111111111	111111111010010	
	0	ro5[51:0]	00000000		0	000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	
	0	ro6[51:0]	00000000			000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	0000000000000111	
	0	ro7[51:0]	00000000		0	000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	0000000000011110	
	0	ro9[51:0]	00000000			000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	0000000000010110	
	0	ro10[51:0]	00000000			000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	000000000000000101	
				1						
				X1	l: 2,199,999 ps					

Fig.7.Simulation results of the 32-point FFT

						2, 199, 999 ps
Name	Value	2,199,995 ps	2,199,996 ps	2,199,997 ps	2,199,998 ps	2,199,999 ps
ro12[51:0] 1	11111111	111	11111111111111111111111	1111111111111111111	111111111101110	
🕨 📑 ro13[51:0] 1	11111111	111	1111111111111111111111	1111111111111111111	111111110111010	
🕨 📑 ro14[51:0] (00000000	000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	000000000101001	
ro15[51:0] 1	11111111	111	11111111111111111111111	111111111111111111111111111111111111111	111111111110110	
ro17[51:0] 1	11111111	111	1111111111111111111111	1111111111111111111	111111111110110	
🕨 📑 ro18[51:0] (0000000	000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	00000000000000100	
🕨 📑 ro19[51:0] 1	11111111	111	1111111111111111111111	1111111111111111111	111111111010001	
▶ 📑 ro20[51:0] 1	11111111	111	1111111111111111111111	1111111111111111111	111111111101110	
🕨 📑 ro21[51:0] (0000000	000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	000000000101000	
ro22[51:0] 1	11111111	111	1111111111111111111111	1111111111111111111	111111111110101	
🕨 📑 ro23[51:0] (00000000	000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	
🕨 📑 ro25[51:0] (00000000	000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	000000000001100	
🕨 📑 ro26[51:0] (0000000	000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	000000000010111	
ro27[51:0]	00000000	000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	
ro28[51:0] 1	11111111	111	1111111111111111111111	1111111111111111111	111111111010010	
🕨 📑 ro29[51:0] (00000000	000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	000000000011110	
🕨 📑 ro30[51:0] 1	11111111	111	111111111111111111111	1111111111111111111	111111111011011	
🕨 📑 ro31[51:0] 1	11111111	111	1111111111111111111111	1111111111111111111	111111111110100	
🕨 📑 io0[53:0]	00000000	0000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	
🕨 📑 io8[53:0]	00000000	0000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	00000000000001110	
🕨 📑 io16[53:0]	0000000	0000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	
▶ 📑 io24[53:0] 1	11111111	1111	1111111111111111111111	111111111111111111111111111111111111111	1111111111110010	
🕨 📑 io4[51:0]	00000000	000	000000000000000000000000000000000000000	000000000000000000000000000000000000000	000000000111001	
▶ 📑 io12[51:0] 1	11111111	111	1111111111111111111111	111111111111111111111111111111111111111	111111111100111	
5 50 - 00 F 4 00		000		000000000000000000000000000000000000000	00000000011001	
		X1: 2,199,999 ps				

Fig.8.Simulation results of the 32-point FFT

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The design summary of the system is as shown in table 1. The design summary shows that the proposed design utilizes 2394 (4%) number of slices registers, 26905 (98%) number of slice LUTs, 1723 (21%) number of fully used LUT-FF pairs, 3107 (1049%) number of bounded IOBs.

The final synthesis report is shown using the table below:

Table 1: Device utilization summary for 32 point FFT using Mixed Radix

Algorithm

Device Utilization Summary for 32 Point FFT using Mixed Radix (Radix-4 & 8) Algorithm					
Logic Utilization	used	Available	Utilization		
No. of Slice Registers	2394	54576	4%		
No. of Slice LUTs	26905	27288	98%		
No. of Fully Used LUT-FF Pairs	1723	8053	21%		
No. of Bonded IOBs	3107	296	1049%		

The timing summary of the proposed Design with Radix-4 &8 Algorithm shows the minimum period is 18.869ns and in Previous Design with Radix-2 Algorithm is 30.783ns. So here the comparison of the proposed design with previous work in terms of slice LUTs, bounded IOBs and time delay is as shown in table 2 and table 3.

Logic utilization	Proposed Design with Radix-4 &8 Algorithm	Previous Design with Radix-2 Algorithm[7]
No. of Slice LUTs	98%	158%
No. of Bonded IOBs	1049%	1706%

Table 2: Comparison of device utilization of 32 point FFT

Table 3: Comparison of Time delay of 32 point FFT

Parameter	Proposed Design with Radix-4 & 8 Algorithm	Previous Design with Radix-2 Algorithm[7]
Minimum Delay(ns)	18.869ns	30.783ns

V. CONCLUSION

In this paper we have designed 32 point FFT design using Radix-4 & 8 algorithm and their simulation and synthesis are done in VHDL using Xilinx synthesis tool on Spartan-6 SP605 Evaluation Platform. The Performance analysis is done by comparing the results of proposed design of 32-point FFT by using Mixed Radix (Radix 4 & 8) with Radix 2 FFT algorithms by taking various parameters into consideration. It is founded that, mixed radix algorithm having better performance over radix-2 algorithm in term of reduction of area complexity and minimum time delay.

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